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Letters

Comments on "Modes of Propagation in a Coaxial Waveguide with Lossless Reactive Guiding Surfaces"

R. A. WALDRON

Many authors have attempted to simplify the study of the mode spectrum of a waveguide with a complicated wall structure by the use of a surface-impedance boundary condition, and the above paper¹ is a classic example of the exercise. The method depends on two assumptions—that it is proper to express a ratio between the tangential E and H fields as a boundary condition, and that such a ratio can be expressed unambiguously in terms of the form of the waveguide wall. That these assumptions are valid is always taken for granted by users of the method, including the present authors, but I have never seen a proof of their validity. Unless such a proof can be given, the results of calculations by the surface-impedance method cannot be trusted.

The results given by the authors in their Fig. 1 appear unusual, and do not agree with the results to the same problem obtained much more simply by applying perturbation theory [1] to the coaxial line, treated as a waveguide [1, sec. IV. F]. This suggests that the assumptions underlying the surface-impedance method require examination. That the method has been widely used does not establish the validity of the assumptions on which it is based.

I have made such a study in [2] where it is shown that there is no value of surface impedance that can be substituted into the characteristic equation obtained by the surface-impedance method that will make it identical with the true characteristic equation. It is also shown that, while approximate agreement between the characteristic equations can be obtained for small surface impedances, the value to be chosen for the surface impedance to secure this agreement is a complicated function of frequency and depends on the mode of propagation. Thus "surface impedance" is not, as has always been sup-

posed, a property of the surface, and its value cannot be known until the solution to the problem under consideration is known. It is therefore of no help in solving a problem. It also follows that the assumption that any desired reactance can be realized is unfounded.

In short, the assumptions on which the surface-impedance method is based are invalid, and it is to this fact that the many strange results can be attributed that have been published by a number of authors. In view of the findings of [2], all results obtained by the surface-impedance method should be treated with caution.

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Reply² by R. K. Arora³

Waldron in his comment, as well as in [2] cited above, has assailed the use of surface impedance as a boundary condition.

The concept of surface impedance is several decades old and its use as a boundary condition has been made by many investigators. The conditions under which a surface may be characterized by an impedance-type boundary condition have been discussed by Senior [1] and Godzinski [2], and a further discussion of the usefulness of these conditions in solving practical problems is available in Barlow and Brown [3]. It is clear from [1] that it is possible, at least in principle, to devise structures with a prescribed value of surface impedance, and so the use of surface impedance as a boundary condition is justifiable on physical grounds. Though it is true that the surface-impedance description is not valid right at the discontinuity, experimental verification is obtained in microwave model experiments for ground-wave propagation [4], [5].

The surface-impedance method has proved to be of value in the solution of many problems of practical interest. No contradictions are

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R. A. Waldron is with the Post Office Research Department, Martlesham Heath, Ipswich, Suffolk, England.

¹ R. K. Arora, S. Vijayaraghavan, and R. Madhavan, *IEEE Trans. Microwave Theory Tech.*, vol. MTT-20, pp. 210-214, Mar. 1972.

² Manuscript received July 25, 1972.

³ R. K. Arora is with the School of Radar Studies, Indian Institute of Technology, Hauz Khas, New Delhi, India.

likely to arise if the results obtained by this method are taken in the proper context. Thus the results reported in the paper under discussion may not be compared with those cited by Waldron for the problem of a coaxial line, as the two problems are different.

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Computer Program Descriptions

Prefabricated Multilayer Section Design Program

PURPOSE: PMSDP searches within the transitional section of an electromagnetic window configuration for the arrangement of prefabricated multilayers that yields the best broad-band frequency-matching condition.

LANGUAGE: Fortran IV; source deck length 250 cards.

AUTHOR: D. L. Huffman, Wright-Patterson Air Force Base, Ohio 45433.

AVAILABILITY: ASIS-NAPS Document No. NAPS-01940.

DESCRIPTION: It is well known that suitably designed transitional sections can reduce the reflection from electromagnetic windows (radomes). The windows treated here are ideal one-dimensional, structures, built up of dielectric materials that are lossless. From the viewpoint of fabricating transitional sections, practical considerations usually dictate that they be composed of N homogeneous layers. In this computer program it is assumed that the N layers have been prefabricated, and the broad-band program is directed toward assembling the N layers in the proper order so that the lowest broad-band reflection coefficient is obtained from the assembled multilayered section. There are $N!$ (factorial) ways of positioning the layers in the multilayered section, and $N!$ multilayered sections are treated.

The order of the dielectric constants and the thicknesses for the prefabricated layers are represented by $DK(I)$ and $TH(I)$, with $I=1, \dots, N$. A portion of the computer program generates the $N!$ permutations of layer position for the N -layered section. Another portion of the program calculates the input reflectivity for NF sampled frequencies. This input reflectivity calculation is repeated until all $N!$ multilayered sections have been considered. All $NF \cdot N!$ input reflectivity values are stored in a two-dimensional matrix. Another portion of the computer program processes the stored input reflectivity matrix data. For each multilayered section, the computer arranges in a descending order of magnitude the values of input reflectivity, that is, the values of input reflectivity that have been calculated for different sampled frequencies. Then the column elements of the processed input reflectivity matrix are listed in ascending value, according to the magnitude of the first row elements as arranged above. Fig. 1 indicates the multilayered transitional section.

APPLICATION

A large number of electromagnetic window configurations can be studied by the permutation search method. The computer program is written for the normal incidence case and only the possibility of a single transitional section is discussed here. Required changes needed to study other general window configurations can easily be provided in the computer program. Experience with the program has indicated that, with $N!$ cases for permutation search available, some transitional sections can be found that produce the desired low reflectivity over a wide frequency range.

Multilayers: $I=1, \dots, N$
 Thicknesses, $TH(I)$
 Relative Dielectric Constants, $DK(I)$
 Load reflectivity, ρ_{load}
 Input reflectivity, ρ_{input}

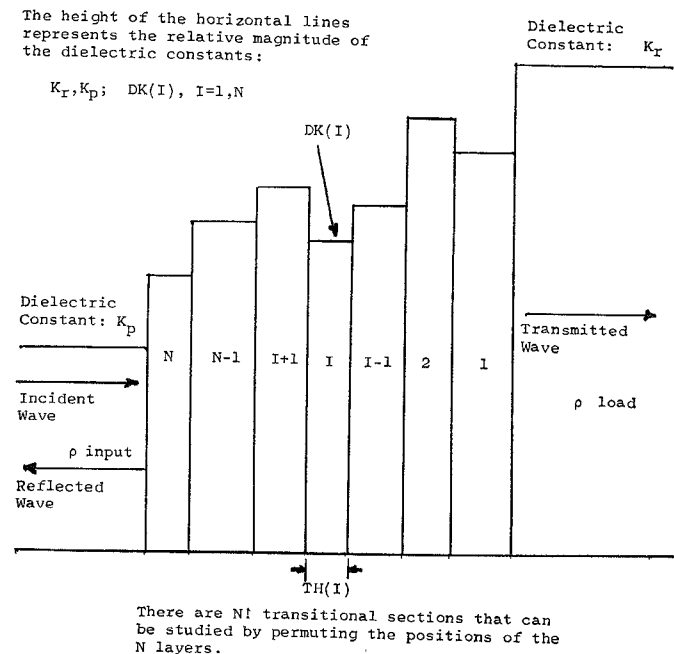


Fig. 1. Geometry of transitional section for an electromagnetic window configuration.

The distinct advantage of this permutation search method is that overall design requirements of space or weight are not involved because any improvement in the broad-band performance can be attributed to the positional arrangement of the prefabricated multilayers. Other optimization methods [1] develop improved broad-band performance by modifying the values for the dielectric constants or the thicknesses of the layers. Such modifications may require an investigation to determine whether the use of the optimum transitional section is feasible.

RESULTS AND COMPUTING TIMES

The basic computer program was executed on Wright-Patterson Air Force Base's CDC 6600 computer. The storage capacity of this computer is 400 000 octal words, thus permitting the chosen number of layers N to be as large as seven. For an electromagnetic window